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Wireless local positioning system

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(12**) United States Patent**

Zekavat

(54) **WIRELESS LOCAL POSITIONING SYSTEM**

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- (51) **Int.Cl.**
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- (52) **U.S. Cl.................. 455/456.1;** 455/456.2; 455/456.5
- (58) **Field of Classification Search** 455/456.1, 455/456.2, 456.3, 456.5, 41.1, 517, 504,

455/506; 342/374,457 See application file for complete search history.

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(57) ABSTRACT

A method of locating mobiles via a dynamic base station. The method includes transmitting a first wireless signal from the dynamic base station at a first time, and receiving the first wireless signal at the mobile. The method also includes transmitting a second wireless signal from the mobile in response to receiving the first wireless signal, and receiving the second wireless signal at the dynamic base station at a second time. The method also includes determining a time difference between the first time and the second time, determining an angle of arrival of the second wireless signal, and locating the mobile based on the angle of arrival and the time difference.

26 Claims, 7 Drawing Sheets

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FIG. 2

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FIG. 7

WIRELESS LOCAL POSITIONING SYSTEM

RELATED APPLICATION

This application claims the benefit of provisional applica-5 tion No. 60/471,553 filed on May 19, 2003, and provisional application No. 60/511,443 filed on Oct. 15, 2003.

BACKGROUND OF THE INVENTION

The present invention relates to communication systems, and more particularly to communication systems configured to locate an object.

SUMMARY OF THE INVENTION

In one form, the invention provides a method of locating mobiles via a dynamic base station. The method includes transmitting a first wireless signal from the dynamic base station at a first time, and receiving the first wireless signal at $_{\,20}$ the mobile. The method also includes transmitting a second wireless signal from the mobile in response to receiving the first wireless signal, and receiving the second wireless signal at the dynamic base station at a second time. The method also includes determining a time difference between the first time 25 and the second time, determining an angle of arrival of the second wireless signal, and locating the mobile based on the angle of arrival and the time difference.

In another form, the invention provides a method of locating a target from a base. The base has an omni-directional or $\,$ 30 $\,$ directional means for transmitting a base wireless signal. The base also includes a directional means such as an antenna array for receiving an target signal and capable of determining a reception angle of the target signal. The target has a transponding means capable of receiving an activating signal 35 and responding with a target signal. The method includes directionally transmitting the activating signal from the omnidirectional means at a first time, activating the transponding means at the target in response to receiving the activating signal, and transmitting a target signal from the transponding 40 means after the transponding means has been activated. The method also includes receiving the target wireless signal at the antenna array means at a second time, determining from the antenna array means the reception angle of the target signal, comparing the first time with the second time to obtain 45 a signal travel time, and locating the target based on the signal travel time and the reception angle of the target signal.

In another form, the invention provides a method of locating a selected one of a plurality of mobiles from a dynamic base. The method includes transmitting a wireless activating 50 signal from the dynamic base at a first time, activating with the wireless activating signal a transponder on each of the plurality of mobiles, in response to activating each transponder, and transmitting with each transponder a wireless signal having a unique mobile signature. The method also includes 55 receiving the wireless signals at the dynamic base at a plurality of arrival times, comparing the unique mobile signature of each wireless signal with a known unique mobile signature of the selected mobile, and identifying the wireless signal of the selected mobile based upon a match between the known 60 unique mobile signal and the unique mobile signal of one of the wireless signals. Furthermore, the method includes determining a reception angle of the wireless signal of the selected mobile, comparing the first time with the arrival time of the wireless signal of the selected mobile to obtain a time differ-65ence, and locating the selected mobile based on the time difference and the reception angle.

Other features and advantages of the invention will become apparent to those skilled in the art upon review of the following detailed description, claims, and drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

In the drawings:

FIG. 1 shows an early warning system according to the invention;

FIG. 2 shows a block diagram of a wireless local positioning system ("WLPS") embodying the invention;

FIG. 3 shows a representation of transmission and reception in the WLPS;

FIG. 4 shows a dynamic base station communicating with ₁₅ a mobile;

FIG. 5 illustrates timing diagrams of signals being received during an interference;

FIG. 6 shows a chart illustrating a probability of overlapping versus a number of transmitter;

FIG. 7 shows a beamformer embodying the invention;

FIG. 8A shows a chart illustrating a probability of detection versus a number of mobiles;

FIG. 8B shows a chart illustrating a probability of detection versus a number of bases;

FIG. 8C shows a magnified chart illustrating a probability of detection versus a number of mobiles of FIG. 8A; and

FIG. 8D shows a magnified chart illustrating a probability of detection versus a number of bases of FIG. 8B.

DETAILED DESCRIPTION

Before any embodiments of the invention are explained in detail, it is to be understood that the invention is not limited in its application to the details of construction and the arrangement of components set forth in the following description or illustrated in the following drawings. The invention is capable of other embodiments and of being practiced or of being carried out in various ways. Also, it is to be understood that the phraseology and terminology used herein is for the purpose of description and should not be regarded as limiting. The use of "including," "comprising," or "having" and variations thereof herein is meant to encompass the items listed thereafter and equivalents thereof as well as additional items. Unless limited otherwise, the terms "connected," "coupled," and "mounted" and variations thereof herein are used broadly and encompass direct and indirect connections, couplings, and mountings. In addition, the terms "connected" and "coupled" and variations thereof are not restricted to physical or mechanical connections or couplings.

FIG. 1 illustrates a vehicle 10 equipped with a vehicle intensified early warning ("VIEW") system 14 according to an embodiment of the invention. The system 14 deploys a secondary surveillance radar system, a secondary radar system, or a dynamic base 18 in the vehicle 10. The radar or the dynamic base system 18 is configured to actively detect for specifically coded transponders or targets such as target 22. A typical target can be a pedestrian carrying a wireless handheld device that includes a specific transponder. For example, a wireless handheld device can be a cell phone, a personal digital assistant ("PDA"), a smart badge worn by children, a specially coded key chain, an animal collar equipped with a transponder, and the like.

Particularly, the radar or the dynamic base system 18 transmits energy bursts consisting of a code specially designed for the vehicle early warning system 14. A transponder of the target 22 detects the code transmitted by a wireless system and automatically responds to the received radar signal. Here,

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the target announces its presence in a space under investigation or a coverage area by the radar or the dynamic base system 18 by sending a burst of energy containing a code (the target's identifier code, e.g., long codes) back to the radar or the dynamic base system 18. The VIEW system 14 also includes other elements such as an antenna 26, a processor 30, and a plurality of sensors 34. The radar or the dynamic base system 18 is configured to be actively sending out detecting signal repetitively. Once the target 22 (that is equipped with a transponder and a unique transponder code) approaching a predetermined distance within a zone of the radar or the dynamic base system 18, the transponder reacts to the sent signal by transmitting the unique transponder code back to the vehicle. The VIEW system 14 and the antenna 26 are also configured to determine the direction in which the target 22 relative to the vehicle 10 with a plurality of signal arrival angles 38, detailed hereinafter. Similarly, a central processing unit 42canalsobe coupled to receive the processed data from the system 14, and to combine the data with the output of a vision warning system 46. The central processing unit 42 will determine if any action is necessary. 20

Hence, each transponder is equipped with a specific transponder code unique for that transponder thereby differentiating one transponder from another. That is, if more than one target is available in the area under investigation, the individual targets can be recognized by the radar system 18. In this way, the early warning system does not need to perform a complex statistical signal processing to detect or recognize the targets in a critical risk of impact and to generate triggering signals. The system 14 equipped with the unique transponder code results in a very high probability of detection (almost 1 in a scale from 0 to 1, with 1 being the highest probability), and a very low probability of false alarm (almost zero in a scale from 0 to 1, with 0 being the lowest probabil- $\frac{35}{25}$ ity).

FIG. 2 shows a wireless local positioning system ("WLPS") 100 block diagram embodying the invention. The WLPS 100 includes two main components, a dynamic base station ("DBS") 104 and a transponder ("TRX") 108 . The $_{40}$ DBS 104, which acts as an dynamic base, is typically deployed in a mobile object such as vehicle, robot, and handheld device such as pocket computer, personal digital assistant, cell phone, and the like. Similarly, the TRX 108 is typically installed in wireless mobile objects or targets such 45 as handheld devices, robots and vehicles, and acts as an active target. In this regard, the vehicle 10 and target 22 described above with respect to FIG. 1 are examples of a DBS 104 and a TRX 108, respectively. However, there are some applications in which one or both of the DBS 104 and the TRX 108 may be stationary rather than mobile, or may sometimes move and sometimes stay still. The DBS 104 transmits periodic ID request ("IDR") signals into a coverage area via an omni-directional antenna 112, when required. Although an omni-directional antenna 112 is shown in FIG. 2, other direc- $_{55}$ tional antenna can also be used in conjunction with the transmitter module 136.

Each TRX 108 is assigned a unique identification ("ID") code. Each TRX 108 also generally includes a simple transceiver that uses a unique signaling transmission and receiving 60 technique, detailed hereinafter. Once the IDR signals generated by the DBS 104 have been detected by a TRX antenna inside the coverage area, the TRX 104 will actively respond to the detected IDR signals. Typically, active responses by the TRX 104 result in a higher probability of detection of the 65uplink signals when compared with signals generated by passive targets. The WLPS 100 also requires minimum or no

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infrastructure operation via the DBS 104 as opposed to a static base station which typically requires communication infrastructures to operate.

Generally, the WLPS 100 is configured to remotely position a TRX 104 within in a line-of-sight ("LOS") of the WLPS 100. In many applications, however, a high coverage area beyond the LOS is required. A high coverage area is created via multi-hop localization techniques. Generally, when an intended TRX is not within a LOS of a searching DBS 104, searching signals from the searching DBS 108 will hop on a second DBS 104'. If the second DBS 104' is within the LOS of the intended TRX 108, the searching DBS 104 is considered an indirect DBS ("IDBS"), whereas the second DBS 104' is considered a direct DBS ("DDBS"). As a result, the IDBS locates a non-LOS TRX via the DDBS that is within a LOS of the intended TRX 108. In this case, the IDBS needs to find a relative location of the DDBS.

Specifically, the TRX 108 includes a transceiver module 116 and an omni-directional antenna 120. The transceiver module 116 typically includes a multiple access transceiver 124 such as a code division multiple access ("CDMA") transceiver, or an ultra-wide band ("UWB") transceiver, and a memory 128. Although the memory 128 is shown being external to the transceiver 124, the memory 128 can also be internal to the transceiver 124 depending on design. In other embodiments, the transceiver module 116 may also include a standard rake receiver 132, and a CDMA transmitter. The TRX 108 transmits a unique ID code as soon as the TRX 108 detects the IDR signals sent by the DBS 104. The unique ID code is assigned permanently at installation, or can be assigned by the DBS 104 depending on application.

The DBS 104, as shown in FIG. 2, includes a DBS transmitter module 136. The DBS transmitter module 136 generates the IDR signal at an IDR signal generator 140 every IDR period or repetition time ("IRT") to all TRX's in its coverage area via a multiple access ("MA") module 144 using MA scheme such as CDMA or UWB-MA, and a modulator 148. FIG. 3 shows a typical timing sequence 300 during a transmission of a IDR signal 302 and reception of an uplink signal 303 from the TRX 108. The IDR signal 302 generally has a signal duration of τ_{DBS} 304, and a period of IRT 308. Between transmissions of the IDR signals 302, the DBS transmitter 136 generally remains idle while a receiver module 156 of the DBS 104 (FIG. 1) waits for a response from any TRX of duration of τ_{TRX} 312. A large IRT 308 generally reduces interference effects at the TRX transceiver module 116 due to a presence of other IDR signals 302 transmitted by other DBS's. The modulated IDR signals or the downlink signal 302 is transmitted via at an omni-directional or directional antenna 112 depending on application.

If an intended TRX 108 is inside the coverage area, the TRX 108 will respond with a unique ID or an uplink signal via its directional or omni-directional antenna 120. The uplink signal 303 generally has a duration of τ_{TRX} that is substantially shorter than the time period between IDR signals—the IRT 308. When the uplink signal 303 is received at a directional antenna array 152, position of each TRX is tracked by the DBS 104 separately. More particularly, a receiver module 156 receives the uplink signal 303 at an MA receiver 160, a diversity combiner 164, an adaptive beamfinder 168, and an angle of arrival ("AOA") finder 172. The operation of the receiver module 156 is detailed hereinafter. The DBS 104 also includes a processor 176 to decode the uplink signal and locate the TRX 108. For collision avoidance applications, the processor 176 also provides a warning signal and controls vehicle brakes when required.

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FIG. **4** shows a general view **400** of how the WLPS **100** finds a TRX position. In general, two factors are used in determining the TRX position, namely, a time of arrival ("TOA") **404,** and direction of arrival ("DOA") or AOA **408.** Once the uplink signal **303** has been received at the MA receiver **60,** the DBS receiver module **156** estimates a roundtrip time, and thus determines a distance **412** between the TRX **108** and the DBS **104.** For example, the distance between the TRX **108** and the DBS **104** can be established by the TOA **404** of the uplink signal **303** with reference to a starting time of the IDR signal **302.** The DOA **408** is determined via the antenna array **152** at the DBS. Specifically, the antenna array **152** uses an AOA detection mechanism. The antenna array **152** also uses a spatial division multiple access ("SDMA") scheme to assist in reducing interference effects. 15 10

Interference effects arise when a plurality of DBS's **104** or a plurality of TRX's **108** are used in the WLPS **100.** The interference effects in the WLPS **100** are mainly due to a probability-of-overlap (" P_{ovl} ") of signals received from the DBS **104** at the TRX transceiver **11**6, or signals received from 20 the TRX **108** at the DBS receiver module **156.** Also, other issue includes range ambiguity at the DBS **104** due to signals transmitted by one TRX **108** in response to IDR signals **302** received from a plurality of DBS's 104. Details of these signal effects are described hereinafter.

In a case where more than one DBS **104** may transmit IDR signals **302** in a specific coverage area, the TRX transceiver **124** is subject to an inter-DBS interference ("IBI"). Selecting a large IRT and using different MA schemes such as direct sequence CDMA ("DS-CDMA") mitigate the IBI effect at 30 the TRX transceiver **124.** Particularly, selecting a large IRT will generally reduce the probability-of-overlap of the received signals from different DBS's at the TRX **108,** thus enhance a probability-of-detection ("POD") of the received signals.

When a number of TRX's **108** in the WLPS coverage area respond to the IDR signal **302** from a DBS **104** simultaneously, an inter-TRX-interference ("IXI") effect at the DBS receiver **156** occurs. FIG. **5** shows two timing diagrams **500, 550** illustrating different types of IXI effects. The timing 40 diagram **500** shows two overlapping ID signals **504,508** from two TRX's in response to an IDR **512** from a DBS **104.** Both ID signals **504, 508** are received within IRT_{min} **516** that are independent of the selection of IRT **520.** Unlike the IBI effects, selecting a large IRT **308** generally does not reduce 45 the IXI effects since the responses from the TRX's **108, 504, 508** are received within a minimum IRT **504** determined by a maximum coverage range of the DBS **104** at all times. To reduce the IXI effects, MA techniques are used for TRX or ID signal transmissions, while directional antenna **152** and 50 beamforming ("BF") techniques are used at the adaptive beamformer **168** of the DBS **104.**

When the TRX **108** uses the same ID signal, or the same MA scheme for IDR's, a range ambiguity occurs. There are a few sources of range ambiguity at the DBS **104.** For example, 55 in a coverage area where more than one DBS is present, each TRX **108** located may generate ID in response to more than one DBS within an IRT. As a result, both the IXI effects and the range ambiguity occur. The timing diagram **550** of FIG. **5** shows three ID signals **554,558,562** from a TRX are received 60 by three different DBS's in response to three different IDR's. Each of the ID signals **554, 558, 562** has a corresponding TOA, **566, 570, 574,** respectively. The three TOA's **566, 570, 574** therefore form a unique range $(R_{554}, R_{558}, R_{562})$ which correspond to **566, 570,** and **574,** respectively. Like the IXI effects are reduced via the MA schemes and the BF techniques as discussed, the range ambiguity is reduced by dif-65

ferent combinations of MA and IDR assignments for each DBS **104.** In this way, a TRX **108** can distinguish a DBS **104** from another DBS **104'** by a unique IDR.

For another example, if an ID signal or a response to an IDR signal **512** is not received by a DBS **104** within the IRT **520,** distances between the DBS **104** and the TRX **108** can be improperly determined. As a result, a DBS **104** generally selects more than one minimum value for the IRT. The minimum value for the IRT is determined by a maximum coverage or a maximum range R_{max} . Considering the maximum uplink antenna array 152 with half power beam widths ("HPBW") β to be less than 90°, the IRT can be determined as follows, which defines a lower limit or a minimum for IRT:

$$
IRT\widetilde{\equiv}IRT_{min}=2T_{max}+T_d+T_G
$$

where T_{max} denotes a maximum possible time delay between a TRX transmission and a DBS reception, T_d is a TRX time delay in responding to the IDR signal, and T_G is a guard band time that corresponds to:

$$
T_G=5T_m+\tau_{DBS}+\tau_{IRX}
$$

where T_m is a channel delay spread, while τ_{DBS} and τ_{TRX} are durations of DBS and TRX transmission signals, respectively. Using simple geometry, T_{max} is determined by R_{max} and β via

$$
T_{\text{max}} = R_{\text{max}} \frac{(1 + \cos^{-1} \beta)}{2c}
$$

where c is the speed of light. The maximum possible time delay of IRT is therefore a function of velocities of the moving TRX and DBS, and the required processing time which varies with applications. In general, the IRT is selected large enough to reduce IBI effects at the TRX receiver. However, a large value of IRT does not affect the IXI, since all of the signals are received by the DBS receiver within T_{max} , which is mainly a function of a maximum coverage range.

The probability-of-overlap has a profound effect on the performance of any TRX or DBS receiver. In general, the probability-of-overlap is a function of the total number of DBS **104** and TRX **108** in the coverage area, and corresponds to

$$
P_{ovl} = 1 - (1 - d_c)^{K-1}
$$

where d_c is a duty cycle, and $d_c = \tau/T$, $\tau = \tau_{DBS}$ or τ_{TRX} is the duration of DBS (TRX) transmitted signal, and $T=T_{DBS}$ or $T=T_{\text{TRX}}=IRT$, and K is the total number of DBS and TRX in the coverage area. In such a case, T_{DBS} =IRT_{min}, as determined earlier. In other embodiments, $\mathrm{T}_{\mathrm{DBS}}$ can be selected based on design and applications. FIG. 6 shows a plurality of curves **600** depicting p_{ovl} **604** as a function of the number of transmitters (TRX or DBS) **608** for different duty cycles **612.** Specifically, as the duty cycle **612** gets smaller, the probability of overlapping decreases. In addition, similar to d_c being a function of τ_{DBS} and τ_{TRX} , performance of both p_{ovl} and the POD of the WLPS 100 also depend on τ_{DBS} and τ_{TRX} . As shown in FIG. 6, selecting a large enough T_{TRX} thus reduces the IBI effects at the TRX transceiver **124.**

In general, selecting values for τ_{DBS} or τ_{TRX} depends on a desired system capacity in terms of the number of TRX **108** and DBS **104** to be accommodated, a system bandwidth and maximum coverage, and applications of the WLPS **100.** In general, both τ_{DBS} and τ_{TRX} are smaller than the IRT to reduce the probability-of-overlapping among signals received by the receivers 124,156 of TRX 108 and DBS 104, respectively. A small P_{ovl} generally decreases both the IBI at the TRX, and the IXI effects at the DBS 108, thereby improves the POD, a positioning accuracy, and a system capacity of the WLPS 100. On the other hand, the system capacity expressed by a 5 maximum number of TRX 108 or DBS 104 determines the number of bits used in each ID code to be transmitted over a period of τ_{DBS} or τ_{TRX} . For a given system capacity, a WLPS bandwidth required for the ID code transmission is inversely proportional to τ_{DBS} or τ_{TRX} A large IRT therefore allows a 10 large τ_{DBS} when compared to τ_{TRX} , without sacrificing P_{ovl} at the TRX transceiver 124. Hence, the WLPS bandwidth is mainly determined by the value of τ_{TRX} .

As discussed earlier, the IXI or IBI effects at the DBS receiver 156 or at the TRX transceiver 124 can be mitigated 15 via selecting a large d_c , and selecting different MA schemes. Furthermore, a large IRT reduces the d_c and consequently the IBI effects at the TRX transceiver 124, while a large IRT does not reduce the IXI effects at the DBS receiver. Flence, while a transceiver 124 or 156 may result in a high TRX POD per-20 formance, performance of the DBS 104 is improved via a combination of MA schemes such as DS-CDMA and SDMA.

In one embodiment, the WLPS 100 includes standard transceivers 136 and 156,116 at the DBS 104 and at the TRX 108, respectively. The transmitted signals from the DBS 104 25 and the TRX 108 are dependent on factors such as a maximum allowable number of bits per ID, and an actual number of bits used in an ID. If a frequency selective channel is used, the uplink or the ID signals received at the DBS 104 are a mixture of signals from different users and different paths. The mix-30 ture of signals can be a combination of a total number of users, a number of paths for the users, a channel fading effect including long-term fading and short term fading, time delay, and a random phase for the user's path, among other things.

As discussed earlier, BF techniques reduce the interference 35 signal from other users and other paths. FIG. 7 shows an exemplary WLPS 700 with a frequency selective channel. The WLPS 700 includes a plurality of antennas 704, and a plurality of rake receivers 708 coupled to a plurality of beamforming paths 712. The signals are first detected by the anten-40 nas 704, the detected wireless signals are fed through the rake receivers 708 for processing. Outputs of the rake receivers are then fed to the beamforming paths 712. Outputs of the beamforming paths are thereafter combined at a combiner 716, and fed to an ID detector 720.

Exemplary applications of the WLPS 100 are illustrated as follows.

In one embodiment, the WLPS 100 is applied to a multirobot cooperation system. In the application, both the DBS 104 and the TRX 108 are mounted on all robots. Multi-robot 50 structures have important applications in space applications, safety systems, and defense technology. In another embodiment, the WLPS 100 includes a collision avoidance system 10. In the collision avoidance system 10, both the TRX 108 vehicle-to-vehicle collision, while pedestrians are required to equipped with a TRX 108.

In yet another embodiment, the WLPS 100 includes an airport security system. In the airport security system, security personnel are equipped with both the DBS 104 and the 60 TRX 108, while other people are required to being equipped with the TRX 108 only. In this way, security personnel will be capable of locating all people in an airport environment. In the application, the DBS 104 may be installed in some positions in hallways, for example, to enhance the coverage area 65beyond the LOS. This is an example of an application in which the DBS 104 is stationary. In addition, multi-hop local -

ization techniques can also be used to extend the coverage area of the DBS 104. In some cases such as indoor applications, UWB-MA systems can be configured to accurately estimate the DOA and the TOA.

In yet another embodiment, the WLPS 100 includes a defense system. The defense system includes a central controller that is equipped with the DBS 104, whereas a commander chains are equipped with both the DBS 104 and the TRX 108, and soldiers are equipped with the TRX 108.

In yet another embodiment, the WLPS 100 is used for law enforcement issues. For example, police vehicles can be equipped with a DBS 104, with a capability of reading the contents of ID codes of each vehicle includes the vehicle make, ownership information, and the like.

Besides vehicle-to-vehicle and vehicle-to-pedestrian collision avoidance, the WLPS 100 has other commercial embodiments. For example, the WLPS 100 can be implemented as a driving aid, monitoring aid, and vehicle office. As a navigational system, the WLPS 100 can help drivers to drive safely and avoid heavy traffic. As a monitoring system, the WLPS 100 is able to collect information instantaneously for emergency help and security controlling.

Specifically, the WLPS 100 can be implemented as a lanemerging-danger-predictor. In either highway or urban streets, most collision accidents happen when driver switches lanes without checking a blind spot behind the driver. The WLPS 100 installed on a vehicle allows the driver to be notified for example with a beep, if there is a vehicle behind the driver closely, the driver will not switch lanes. In this application, directional antennas 112 are used with the DBS 104 transmitter 136 for specific coverage areas.

Another application of the WLPS 100 is automatic traffic sign notification. In urban area, the TRX can be installed on stop signs, street sign, one way street sign, yield sign, gas station, and the like, the driver can be alerted of the signs ahead automatically, for example, by voice and/or text.

Also, on interstate highways, the TRX 108 can be installed on exit signs. In this way, the driver will not have to miss a particular exit. The driver can even set a vehicle destination exit, and the driver will be notified when the destination exit is near. Furthermore, installing the TRX 108 on the retroreflective pavement markers used to guide drivers on highway, an auto-pilot system can drive the vehicle via these ₄₅ guiding signals.

In addition to the driving aid, the WLPS 100 can also be employed in heavy traffic controls. For example, a parking lot of a stadium can install a unique TRX 108. When the lot is full, the TRX 108 will transmit a "full" signal, and the drivers will be notified. In general, the application of WLPS 100 in driving is to "PUSH" the information to drivers, and then the driver does not need to "PULL" the information by the operator.

and the DBS 104 are mounted on all vehicles to prevent 55 alternative ID of a vehicle. The transmitted signal of a TRX As for administration, the WLPS 100 can be used as an 108 on the vehicle can include the vehicle's maker, year, color, VIN, plate number, owner's name, and the like. Therefore, a TRX can be used as an ID of the vehicle. The transmitted ID is generally encrypted to ensure privacy. Considering that access time to each ID is quite short, it is very time consuming to decrypt the ID via brutal decoding. Only authorized people such as police cars may have privileges to read the contents of the ID. All vehicle information including position, speeds and brief information, are shown in a small screen. When accidents happen, a policeman and an ambulance can retrieve the vehicle's information and location before arriving.

Furthermore, the WLPS **100** can also be applied to automatic toll payment. In cities where vehicles have to stop by some toll booths to pay tolls which slows down the traffic considerably in many cases. With the WLPS **100,** the vehicles bypass these stations, and a charging process is taken auto-5 matically when they pass through. In this way, the drivers do not need to carry any coin or cash, the stations save their human-power cost, and the traffic becomes smooth. In this example, the toll booth acts as a stationary DBS.

The WLPS **100** can be considered as an add-on to a vehicle. 10 For example, the WLPS **100** is a part of a vehicle. The WLPS 100 will co-operate with a vehicle self-diagnostic system, a GPS system, a mobile communication system, an auto-pilot system, and the like. Also, considering each vehicle as a node of a wireless sensor network, the vehicles can exchange their information. By implementing the WLPS **100,** the drivers are able to use their vehicle as an office. The police can monitor the traffic, trace the stolen car, locate an accident, stop a heavily damaged car, etc. In general, the WLPS **100** helps the information exchange between drivers, polices and the road 20 utilities in the future vehicles.

In one embodiment, the WLPS system includes a plurality of TRX's, a multi-path environment. The ID code has six bits, the DS-CDMA code has 64 chips, the channel delay spread for a typical street area is 27 η sec, and the carrier frequency is $\,$ 25 $\,$ 3 GHz, τ_{DBS} is 24 µsec, τ_{TRX} is 1.2 µsec. Other features include a maximum coverage range of 1000 m, the antenna array being linear with four elements with antenna element separation of 0.05 m and HPBW being 27°, and four multipaths leading to four fold path diversity. Furthermore, the TRX distance is uniformly distributed in [0 1] km, the TRX angle is uniformly distributed in $[0 \pi]$, the TRX TOA is uniformly distributed in $[0 T_{max}]$ with T_{max} being 3.3 µsec, a uniform multi-path intensity profile, i.e., bit energy is distributed in each path identically, binary phase shift keying 35 ("BPSK") modulation, and perfect power control and TOA estimation. Based on the setup, the minimum IRT, IRT_{min} is 9.83 usec. A value of IRT=24 msec is also selected to reduce the IBI effects. With the τ_{DBS} and τ_{TRX} , the bandwidth of a DS-CDMA (standard) transmitter is 320 MHz (5 MHz) for 40 the TRX, 16 MFiz (250 KHz) for the DBS, which is much smaller than the TRX bandwidth. Hence, the WLPS bandwidth is mainly determined by the TRX transmission bandwidth, as expected. In addition, using these parameters, the duty cycle for the DBS and the TRX receivers correspond to 45 d_{cDBS} =0.1 and d_{cTRX} =0.001. In some cases, the WLPS **100** has a bandwidth between about 2 GHz and 3 GHz, or between 100 MHz and 10 GHz depending on application. 30

As discussed earlier, the IBI effects at the TRX receiver can be considerably reduced by selecting a large IRT. However, 50 the selection will not affect the IXI effects at the DBS receiver **156.** Hence, a TRX transceiver module **116** can be implemented by the transceiver **124,** or by using a DS-CDMA scheme without BF. The DBS receiver **156,** however, needs a combination of DS-CDMA and BF. A small $d_{c,TRX}=0.001$ at 55 the TRX transceiver 124 leads to a small p_{ovl} , which leads to small IBI effects and high POD. In contrast, a large $d_{c,DBS}$ being about 0.1 at the DBS receiver leads to a high p_{out} that results in high IXI effects. Both BF and CDMA techniques help to reduce the IXI effects at the DBS.

The probability of detection of the DBS receiver is shown in FIG. 8A which shows the POD versus a number of TRX's for a standard transceiver **124** and a DS-CDMA transceiver **132,** without antenna arrays **152** and BF **168.** FIG. 8A also shows that in general the POD decreases as the number of 65TRX's increases, which is a direct result of the IXI effects. As to the impact of BF, the use of BF does not affect much the

capacity in terms of number of TRX's for a standard receiver, but it considerably enhances the capacity of the DS-CDMA system. Merging DS-CDMA with BF thus enhances the POD performance of the WLPS system.

The POD results for a TRX transceiver **124** using a standard receiver is shown in FIG. 8B which shows that a standard TRX receiver achieves good POD performance, although the POD performance can be improved if a larger IRT value, or a smaller τ_{DBS} value is used. For example, occupying the same bandwidth as DS-CDMA, a standard receiver has τ_{DBS} or τ_{TRX} to be 1/64th of that of a DS-CDMA system. In such a case, the same number of path diversity as the DS-CDMA receiver, that is, four fold diversity, can be achieved. In some cases, the system has a bandwidth between about 2 GHz and 3 GHz, or between 100 MHz and 10 GHz.

FIG. 8B shows that, with similar bandwidths, a DS-CDMA system with duty cycle 64 times higher than a standard receiver leads to almost a same performance curve. The top two curves in FIG. 8A for standard and DS-CDMA receivers with BF are magnified in FIG. 8C. FIG. 8C shows that standard receiver outperforms DS-CDMA receiver for DS-CDMA with high duty cycles. In this case, the standard receiver leads to a capacity more than two times of DS-CDMA receiver at the POD=0.99. For the TRX transceiver **124,** p_{ov} statistics with d_c about 0.000015 is depicted. The low probability leads to the high TRX standard receiver POD performance shown in FIG. 8B. However, FIG. 8B also shows that DS-CDMA receiver outperforms standard receiver for DS-CDMA with low duty cycles. For airport security where the data rate is not critical, the IRT can be chosen larger than 24 msec for even better TRX standard receiver POD performance.

Various features and advantages of the invention are set forth in the following claims.

The invention claimed is:

1. A method of locating an immobile target fixedly positioned at a location via a mobile base, the method comprising: transmitting a first wireless signal from the mobile base at a first time;

receiving the first wireless signal at the immobile target fixedly positioned at the location;

transmitting a second wireless signal from the immobile target in response to receiving the first wireless signal;

- receiving the second wireless signal at the mobile base at a second time;
- determining a time difference between the first time and the second time;
- determining an angle of arrival of the second wireless signal; and
- locating the immobile target based on the angle of arrival and the time difference.
- 2. The method of claim I, wherein transmitting the first wireless signal comprises:

generating an identification request; and

modulating the identification request with a multiple access scheme.

3. The method of claim 2, wherein the multiple access scheme comprises at least one of a code-division multiple access ("CDMA") scheme, a direct sequence CD MA ("DS-CDMA"), a synchronous CDMA ("SCOMA"), and an ultra-60 wide band multiple access ("UWB-MA").

4. The method of claim **1,** wherein transmitting the first wireless signal comprises providing a carrier frequency between about 2 GHz and about 3 GHz.

5. The method of claim **1,** wherein receiving the second wireless signal comprises receiving the second wireless signal at least one of an antenna array and a rake receiver array.

6. The method of claim 1, wherein transmitting the first wireless signal comprises transmitting the first wireless signal using at least one of a long term fading technique, and short term fading technique.

7. The method of claim 1, further comprising providing a 5 communication bandwidth between about 10 MHz and about 10 GHz.

8. The method of claim 1, further comprising beamforming the second wireless signal.

9. The method of claim 1, wherein transmitting the first 10 wireless signal comprises omni-directionally transmitting the first wireless signal.

10. The method of claim 1, wherein determining a time difference comprises determining a velocity of the mobile base. 15

11. A method of locating an immobile target fixedly positioned at a location from a mobile base, wherein the mobile base has an omni-directional means for transmitting a base wireless signal, and an antenna array means for receiving a target signal and capable of determining a reception angle of 20 the target signal, and the target at the fixed location has a transponding means capable of receiving an activating signal and responding with a target signal, the method comprising:

- omni-directionally transmitting the activating signal from the omni-directional means at a first time while moving 25 the mobile base;
- activating the transponding means at the immobile target fixedly positioned at the location in response to receiving the activating signal;
- transmitting a wireless target signal from the transponding 30 means after the transponding means has been activated;
- receiving the target signal at the antenna array means at a second time;
- determining from the antenna array means the reception angle of the target signal;
- comparing the first time with the second time to obtain a signal travel time; and
- locating the immobile target based on the signal travel time and the reception angle of the target signal.

12. The method of claim 11, wherein transmitting the acti-40 vating signal comprises:

generating an identification request; and

modulating the identification request with a multiple access scheme.

13. The method of claim 12, wherein the multiple access 45 scheme comprises at least one of a code-division multiple access ("CDMA") scheme, a direct sequence CDMA ("DS-CDMA"), a synchronous CDMA ("CDMA"), and an ultrawide band multiple access ("UWB-MA").

14. The method of claim 13, wherein transmitting the acti-50 vating signal comprises providing a carrier frequency between about 2 and about 3 GHz.

15. The method of claim 11, wherein receiving the target wireless signal comprises receiving the target wireless signal at a rake receiver array. 55

16. The method of claim 11, wherein transmitting the activating signal comprises transmitting the activating signal using at least one of a long term fading technique, and short term fading technique.

17. The method of claim 11, further comprising providing a communication bandwidth between about 10 MHz and 10 GHz.

18. The method of claim 11, further comprising beamforming the target wireless signal.

19. A method of locating a selected one of a plurality of immobile targets fixedly positioned at respective locations from a mobile base, the method comprising:

- transmitting a wireless activating signal from the mobile base at a first time;
- activating with the wireless activating signal a transponder on each of the plurality of immobile targets at their respective fixed locations;
- in response to activating each transponder, transmitting with each transponder at its fixed location a wireless signal having a unique mobile signature;
- receiving the wireless signals at the mobile base at a plurality of arrival times;
- comparing the unique mobile signature of each wireless signal with a known unique mobile signature of the selected immobile target;
- identifying the wireless signal of the selected immobile target based upon a match between the known unique mobile signal and the unique mobile signal of one of the wireless signals;
- determining a reception angle of the wireless signal of the selected immobile target;
- comparing the first time with the arrival time of the wireless signal of the selected immobile target to obtain a time difference; and
- locating the selected immobile target based on the time difference and the reception angle.

20. The method of claim 19, wherein transmitting the wireless activating signal comprises:

generating an identification request; and

modulating the identification request with a multiple access scheme.

21. The method of claim 20, wherein the multiple access scheme comprises at least one of a code-division multiple access ("CDMA") scheme, a direct sequence CDMA ("DS-CDMA"), a synchronous CDMA ("SCDMA"), and an ultra-

wide band multiple access ("UWB-MA").

22. The method of claim 19, wherein transmitting the wireless activating signal comprises providing a carrier frequency between about 2 GHz and about 3 GHz.

23. The method of claim 19, wherein receiving the mobile wireless signal comprises receiving the mobile wireless signal at a rake receiver array.

24. The method of claim 19, wherein transmitting the wireless activating signal comprises transmitting the wireless activating signal using at least one of a long term fading technique, and short term fading technique.

25. The method of claim 19, further comprising providing a communication bandwidth between about 10 MHz and 10 GHz.

26. The method of claim 19, further comprising beamforming the mobile wireless signal.

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